

CHAPTER 9

P.P.9.1 amplitude = 5
 phase = -60°
 angular frequency (ω) = 4π = 12.57 rad/s
 period (T) = $\frac{2\pi}{\omega}$ = 0.5 s
 frequency (f) = $\frac{1}{T}$ = 2 Hz

P.P.9.2 $i_1 = -4\sin(\omega t + 25^\circ) = 4\cos(\omega t + 25^\circ + 90^\circ)$
 $i_1 = 4\cos(\omega t + 115^\circ)$, $\omega = 377 \text{ rad/s}$

Compare this with

$$i_2 = 5\cos(\omega t - 40^\circ)$$

indicates that the phase angle between i_1 and i_2 is

$$115^\circ + 40^\circ = 155^\circ$$

Thus, **i_1 leads i_2 by 155°**

P.P.9.3 (a) $(5 + j2)(-1 + j4) = -5 + j20 - j2 - 8 = -13 + j18$
 $5\angle 60^\circ = 2.5 + j4.33$
 $(5 + j2)(-1 + j4) - 5\angle 60^\circ = -15.5 + j13.67$
 $[(5 + j2)(-1 + j4) - 5\angle 60^\circ]^* = \underline{-15.5 - j13.67} = \underline{20.67\angle 221.41^\circ}$

(b) $3\angle 40^\circ = 2.298 + j1.928$
 $10 + j5 + 3\angle 40^\circ = 12.298 + j6.928 = 14.115\angle 29.39^\circ$
 $-3 + j4 = 5\angle 126.87^\circ$
 $\frac{10 + j5 + 3\angle 40^\circ}{-3 + j4} = \frac{14.115\angle 29.39^\circ}{5\angle 126.87^\circ} = 2.823\angle -97.48^\circ$
 $2.823\angle -97.48^\circ = -0.3675 - j2.8$
 $10\angle 30^\circ = 8.66 + j5$
 $\frac{10 + j5 + 3\angle 40^\circ}{-3 + j4} + 10\angle 30^\circ = \underline{8.293 + j2.2}$

P.P.9.4 (a) $-\cos(A) = \cos(A - 180^\circ) = \cos(A + 180^\circ)$
 Hence,
 $v = -7\cos(2t + 40^\circ) = 7\cos(2t + 40^\circ + 180^\circ)$
 $v = 7\cos(2t + 220^\circ)$
 The phasor form is
 $V = 7\angle 220^\circ \text{ V}$
 (b) Since $\sin(A) = \cos(A - 90^\circ)$,

$$i = 4 \sin(10t + 10^\circ) = 4 \cos(10t + 10^\circ - 90^\circ)$$

$$i = 4 \cos(10t - 80^\circ)$$

The phasor form is

$$\underline{I} = 4\angle -80^\circ \text{ A}$$

P.P.9.5

(a) Since $-1 = 1\angle -180^\circ = 1\angle 180^\circ$

$$\underline{V} = -10\angle 30^\circ = 10\angle (30^\circ + 180^\circ) = 10\angle 210^\circ$$

The sinusoid is

$$v(t) = \underline{10 \cos(\omega t + 210^\circ) \text{ V}}$$

(b) $\underline{I} = j(5 - j12) = 12 + j5 = 13\angle 22.62^\circ$

The sinusoid is

$$\underline{i(t)} = \underline{13 \cos(\omega t + 22.62^\circ) \text{ A}}$$

P.P.9.6

Let $V = -10\sin(\omega t + 30^\circ) + 20\cos(\omega t - 45^\circ)$

Then, $V = 10\cos(\omega t + 30^\circ + 90^\circ) + 20\cos(\omega t - 45^\circ)$

Taking the phasor of each term

$$\underline{V} = 10\angle 120^\circ + 20\angle -45^\circ$$

$$\underline{V} = -5 + j8.66 + 14.14 - j14.14$$

$$\underline{V} = 9.14 - j5.48 = 10.66\angle -30.95^\circ$$

Converting $\underline{V}$ to the time domain

$$v(t) = \underline{10.66 \cos(\omega t - 30.95^\circ) \text{ V}}$$

P.P.9.7

Given that

$$2\frac{dv}{dt} + 5v + 10\int v dt = 20\cos(5t - 30^\circ)$$

we take the phasor of each term to get

$$2j\omega \underline{V} + 5 \underline{V} + \frac{10}{j\omega} \underline{V} = 20\angle -30^\circ, \quad \omega = 5$$

$$\underline{V} [j10 + 5 - j(10/5)] = \underline{V} (5 + j8) = 20\angle -30^\circ$$

$$\underline{V} = \frac{20\angle -30^\circ}{5 + j8} = \frac{20\angle -30^\circ}{9.434\angle 58^\circ}$$

$$\underline{V} = 2.12\angle -88^\circ$$

Converting $\underline{V}$ to the time domain

$$v(t) = \underline{2.12 \cos(5t - 88^\circ) \text{ V}}$$

P.P.9.8

For the capacitor,

$$\underline{V} = \underline{I} / (j\omega C), \text{ where } \underline{V} = 6\angle -30^\circ, \quad \omega = 100$$

$$\underline{I} = j\omega C \underline{V} = (j100)(50 \times 10^{-6})(6\angle -30^\circ)$$

$$\underline{I} = 30\angle 60^\circ \text{ mA}$$

$$i(t) = \underline{30 \cos(100t + 60^\circ) \text{ mA}}$$

P.P.9.9

$$\underline{V_s} = 5\angle 0^\circ, \quad \omega = 10$$

$$\mathbf{Z} = 4 + j\omega L = 4 + j2$$

$$\mathbf{I} = \mathbf{V}_s / \mathbf{Z} = \frac{5\angle 0^\circ}{4 + j2} = \frac{5(4 - j2)}{16 + 4} = 1 - j0.5 = 1.118\angle -26.57^\circ$$

$$\mathbf{V} = j\omega L \mathbf{I} = j2 \mathbf{I} = (2\angle 90^\circ)(1.118\angle -26.57^\circ) = 2.236\angle 63.43^\circ$$

Therefore, $v(t) = \underline{\underline{2.236 \sin(10t + 63.43^\circ) \text{ V}}}$
 $i(t) = \underline{\underline{1.118 \sin(10t - 26.57^\circ) \text{ A}}}$

P.P.9.10

Let $\mathbf{Z}_1$ = impedance of the 2-mF capacitor in series with the 20- Ω resistor

$\mathbf{Z}_2$ = impedance of the 4-mF capacitor

$\mathbf{Z}_3$ = impedance of the 2-H inductor in series with the 50- Ω resistor

$$\mathbf{Z}_1 = 20 + \frac{1}{j\omega C} = 20 + \frac{1}{j(10)(2 \times 10^{-3})} = 20 - j50$$

$$\mathbf{Z}_2 = \frac{1}{j\omega C} = \frac{1}{j(10)(4 \times 10^{-3})} = -j25$$

$$\mathbf{Z}_3 = 50 + j\omega L = 50 + j(10)(2) = 50 + j20$$

$$\mathbf{Z}_{in} = \mathbf{Z}_1 + \mathbf{Z}_2 \parallel \mathbf{Z}_3 = \mathbf{Z}_1 + \mathbf{Z}_2 \mathbf{Z}_3 / (\mathbf{Z}_2 + \mathbf{Z}_3)$$

$$\mathbf{Z}_{in} = 20 - j50 + \frac{-j25(50 + j20)}{-j25 + 50 + j20}$$

$$\mathbf{Z}_{in} = 20 - j50 + 12.38 - j23.76$$

$$\mathbf{Z}_{in} = \underline{\underline{32.38 - j73.76 \Omega}}$$

P.P.9.11 In the frequency domain,

the voltage source is $\mathbf{V}_s = 10\angle 75^\circ$

the 0.5-H inductor is $j\omega L = j(10)(0.5) = j5$

the $\frac{1}{20}$ -F capacitor is $\frac{1}{j\omega C} = \frac{1}{j(10)(1/20)} = -j2$

Let $\mathbf{Z}_1$ = impedance of the 0.5-H inductor in parallel with the 10- Ω resistor

and $\mathbf{Z}_2$ = impedance of the (1/20)-F capacitor

$$\mathbf{Z}_1 = 10 \parallel j5 = \frac{(10)(j5)}{10 + j5} = 2 + j4 \quad \text{and} \quad \mathbf{Z}_2 = -j2$$

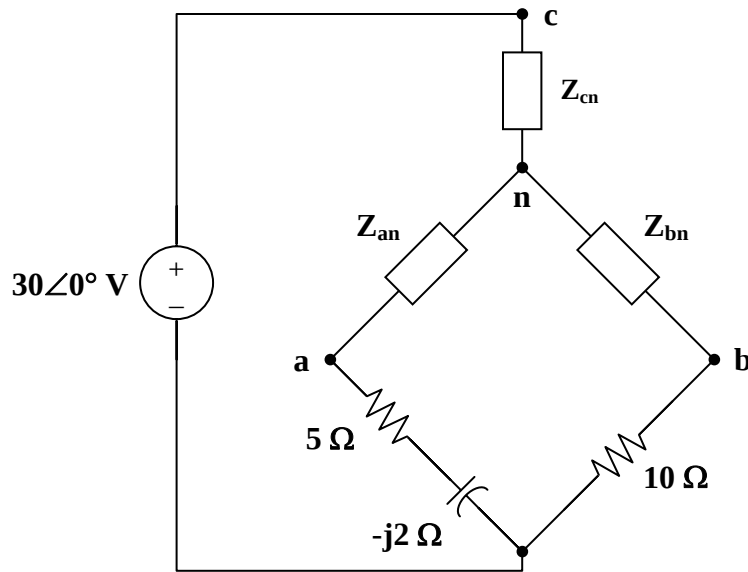
$$\mathbf{V}_o = \mathbf{Z}_2 / (\mathbf{Z}_1 + \mathbf{Z}_2) \mathbf{V}_s$$

$$\mathbf{V}_o = \frac{-j2}{2 + j4 - j2} (10\angle 75^\circ) = \frac{-j(10\angle 75^\circ)}{1 + j} = \frac{10\angle (75^\circ - 90^\circ)}{\sqrt{2}\angle 45^\circ}$$

$$\mathbf{V}_o = 7.071\angle -60^\circ$$

$$v_o(t) = \underline{\underline{7.071 \cos(10t - 60^\circ) \text{ V}}}$$

P.P.9.12 We need to find the equivalent impedance via a delta-to-wye transformation as shown below.



$$Z_{an} = \frac{j4(8 + j5)}{j4 + 8 + j5 - j3} = \frac{4(-5 + j8)}{8 + j6} = 0.32 + j3.76$$

$$Z_{bn} = \frac{-j3(8 + j5)}{8 + j6} = \frac{3(5 - j8)(8 - j6)}{100} = -0.24 - j2.82$$

$$Z_{cn} = \frac{j4(-j3)}{8 + j6} = \frac{12(8 - j6)}{100} = 0.96 - j0.72$$

The total impedance from the source terminals is

$$Z = Z_{cn} + (Z_{an} + 5 - j2) \parallel (Z_{bn} + 10)$$

$$Z = Z_{cn} + (5.32 + j1.76) \parallel (9.76 - j2.82)$$

$$Z = Z_{cn} + \frac{(5.32 + j1.76)(9.76 - j2.82)}{(5.32 + j1.76) + (9.76 - j2.82)}$$

$$Z = 0.96 - j0.72 + 3.744 + j0.4074$$

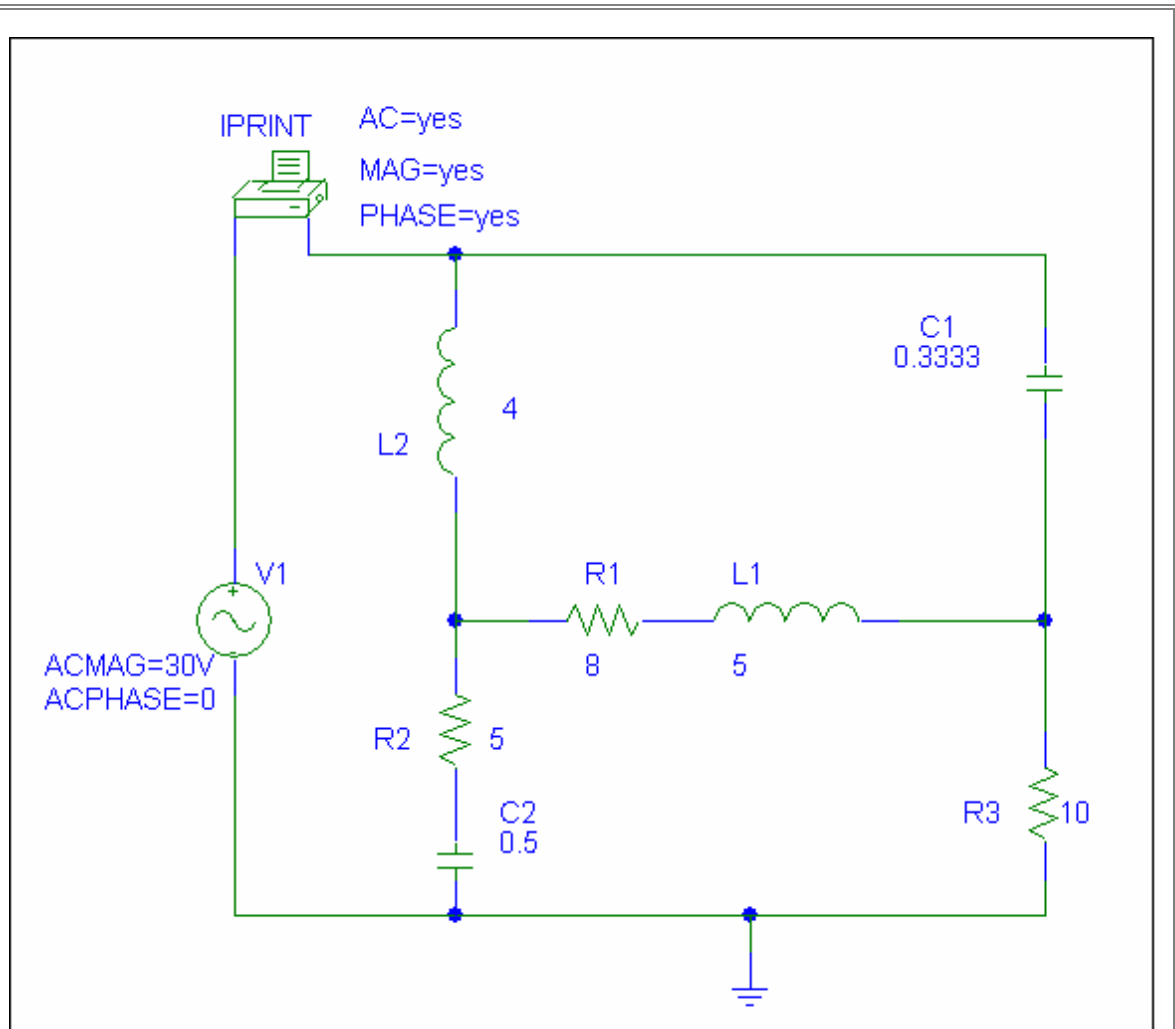
$$Z = 4.704 - j0.3126 = 4.714 \angle -3.802^\circ$$

Therefore,

$$I = V / Z = \frac{30 \angle 0^\circ}{4.714 \angle -3.802^\circ}$$

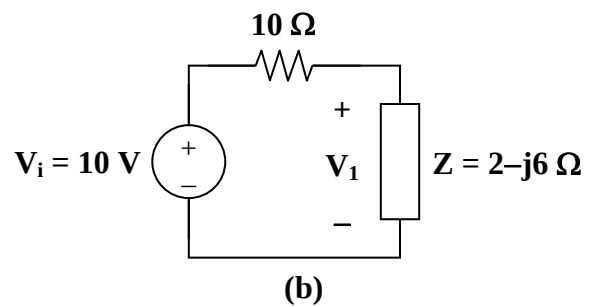
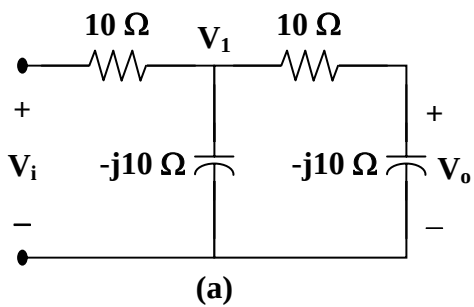
$$I = \underline{\underline{6.364 \angle 3.802^\circ \text{ A}}}$$

Let us now check this using PSpice. The solution produces the magnitude of $I = 6.364\text{E}+00$, and the phase angle = $3.803\text{E}+00$, which agrees with the above answer.



P.P.9.13 To show that the circuit in Fig. (a) meets the requirement, consider the equivalent circuit in Fig. (b).

$$\mathbf{Z} = -j10 \parallel (10 - j10) = \frac{-j10(10 - j10)}{10 - j20} = \frac{-j(10 - j10)}{1 - j2} = 2 - j6 \, \Omega$$



$$\mathbf{V}_1 = \frac{2 - j6}{10 + 2 - j6} (10) = \frac{10}{3} (1 - j)$$

$$\mathbf{V}_o = \frac{-j10}{10 - j10} \mathbf{V}_1 = \left(\frac{-j}{1 - j} \right) \left(\frac{10}{3} \right) (1 - j) = -j \frac{10}{3}$$

$$\mathbf{V}_o = \frac{10}{3} \angle -90^\circ$$

This implies that the RC circuit provides a 90° lagging phase shift.

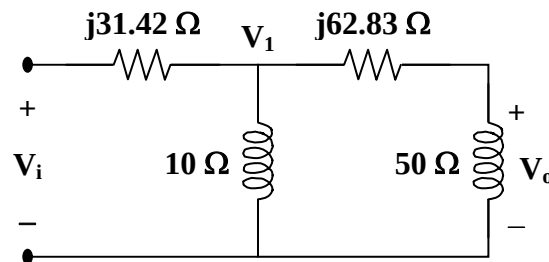
The output voltage is $\frac{10}{3} = \underline{\underline{3.333 \text{ V}}}$

P.P.9.14

the 1-mH inductor is $j\omega L = j(2\pi)(5 \times 10^3)(1 \times 10^{-3}) = j31.42$

the 2-mH inductor is $j\omega L = j(2\pi)(5 \times 10^3)(2 \times 10^{-3}) = j62.83$

Consider the circuit shown below.



$$\mathbf{Z} = 10 \parallel (50 + j62.83) = \frac{(10)(50 + j62.83)}{60 + j62.83}$$

$$\mathbf{Z} = 9.205 + j0.833 = 9.243 \angle 5.17^\circ$$

$$\mathbf{V}_1 = \mathbf{Z} / (\mathbf{Z} + j31.42) \mathbf{V}_i = \frac{9.243 \angle 5.17^\circ}{9.205 + j32.253} (1) = 0.276 \angle -68.9^\circ$$

$$\mathbf{V}_o = \frac{50}{50 + j62.83} \mathbf{V}_1 = \frac{50(0.276 \angle -68.9^\circ)}{80.297 \angle 51.49^\circ} = 0.172 \angle -120.4^\circ$$

Therefore,

magnitude = **0.172**

phase = **120.4°**

phase shift is **lagging**

P.P.9.15 $\mathbf{Z_x = (Z_3 / Z_1) Z_2}$

$$\mathbf{Z_3 = 12\ k\Omega}$$

$$\mathbf{Z_1 = 4.8\ k\Omega}$$

$$\mathbf{Z_2 = 10 + j\omega L = 10 + j(2\pi)(6 \times 10^6)(0.25 \times 10^{-6}) = 10 + j9.425}$$

$$\mathbf{Z_x = \frac{12k}{4.8k}(10 + j9.425) = 25 + j23.5625\ \Omega}$$

$$\mathbf{R_x = 25, \quad X_x = 23.5625 = \omega L_x}$$

$$\mathbf{L_x = \frac{X_x}{2\pi f} = \frac{23.5625}{2\pi(6 \times 10^6)} = 0.625\ \mu\text{H}}$$

i.e. **a 25- Ω resistor in series with a 0.625- μH inductor.**